SA676

Low-voltage mixer FM IF system

Rev. 3 — 19 July 2012 Product data sheet

1. General description

The SA676 is a low-voltage monolithic FM IF system incorporating a mixer/oscillator, two limiting intermediate frequency amplifiers, quadrature detector, logarithmic Received Signal Strength Indicator (RSSI), voltage regulator and audio and RSSI op amps. The SA676 is available in a 20-pin SSOP (Shrink Small Outline Package).

The SA676 was designed for cordless telephone applications in which efficient and economic integrated solutions are required and yet high performance is desirable. Although the product is not targeted to meet the stringent specifications of high performance cellular equipment, it will exceed the needs for analog cordless phones. The minimal amount of external components and absence of any external adjustments makes for a very economical solution.

2. Features and benefits

- Low power consumption: 3.5 mA typical at 3 V
- Mixer input to > 100 MHz
- Mixer conversion power gain of 17 dB at 45 MHz
- XTAL oscillator effective to 100 MHz (LC oscillator or external oscillator can be used at higher frequencies)
- 102 dB of IF amplifier/limiter gain
- 2 MHz IF amp/limiter small signal bandwidth
- Temperature compensated logarithmic Received Signal Strength Indicator (RSSI) with a 70 dB dynamic range
- Low external component count; suitable for crystal/ceramic/LC filters
- Audio output internal op amp
- RSSI output internal op amp
- Internal op amps with rail-to-rail outputs
- ESD protection exceeds 2000 V HBM per JESD22-A114 and 1000 V CDM per JESD22-C101
- Latch-up testing is done to JEDEC Standard JESD78 Class II, Level B

3. Applications

- Cordless telephones

4. Ordering information

Table 1. Ordering information

T_{amb} = −40 °C to +85 °C

Type number	Topside mark	Package		
		Name	Description	Version
SA676DK/01	SA676DK	SSOP20	plastic shrink small outline package; 20 leads; body width 4.4 mm	SOT266-1

5. Block diagram

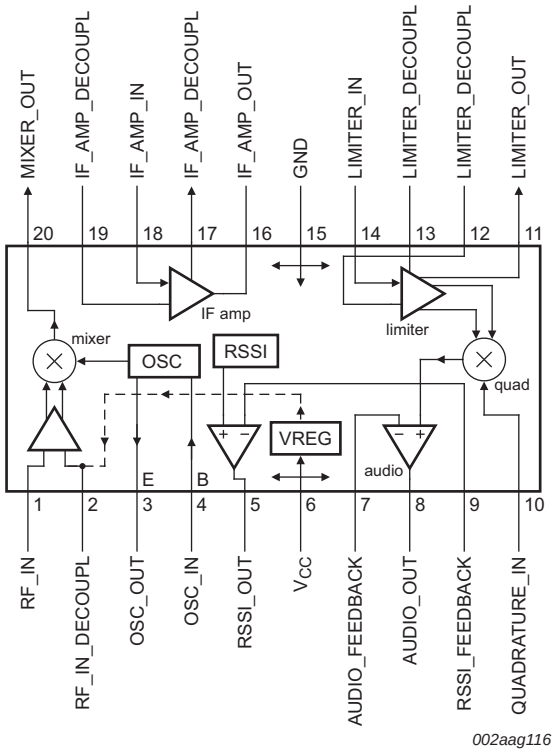


Fig 1. Block diagram

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6. Pinning information

6.1 Pinning

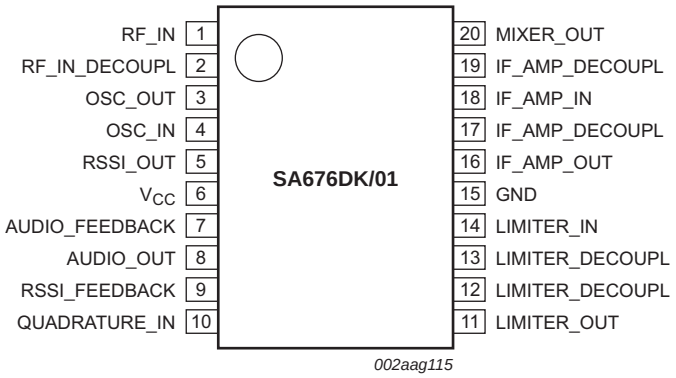


Fig 2. Pin configuration for SSOP20

6.2 Pin description

Table 2. Pin description

Symbol	Pin	Description
RF_IN	1	RF input
RF_IN_DECOUPL	2	RF input decoupling pin
OSC_OUT	3	oscillator output
OSC_IN	4	oscillator input
RSSI_OUT	5	RSSI output
V_{CC}	6	positive supply voltage
AUDIO_FEEDBACK	7	audio amplifier negative feedback terminal
AUDIO_OUT	8	audio amplifier output
RSSI_FEEDBACK	9	RSSI amplifier negative feedback terminal
QUADRATURE_IN	10	quadrature detector input terminal
LIMITER_OUT	11	limiter amplifier output
LIMITER_DECOUPL	12	limiter amplifier decoupling pin
LIMITER_DECOUPL	13	limiter amplifier decoupling pin
LIMITER_IN	14	limiter amplifier input
GND	15	ground; negative supply
IF_AMP_OUT	16	IF amplifier output
IF_AMP_DECOUPL	17	IF amplifier decoupling pin
IF_AMP_IN	18	IF amplifier input
IF_AMP_DECOUPL	19	IF amplifier decoupling pin
MIXER_OUT	20	mixer output

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7. Functional description

The SA676 is an IF signal processing system suitable for second IF systems with input frequency as high as 100 MHz. The bandwidth of the IF amplifier and limiter is at least 2 MHz with 90 dB of gain. The gain/bandwidth distribution is optimized for 455 kHz, 1.5 kΩ source applications. The overall system is well-suited to battery operation as well as high performance and high quality products of all types.

The input stage is a Gilbert cell mixer with oscillator. Typical mixer characteristics include a noise figure of 7.0 dB, conversion gain of 17 dB, and input third-order intercept of −10 dBm. The oscillator will operate in excess of 100 MHz in L/C tank configurations. Hartley or Colpitts circuits can be used up to 100 MHz for crystal configurations.

The output impedance of the mixer is a 1.5 kΩ resistor permitting direct connection to a 455 kHz ceramic filter. The input resistance of the limiting IF amplifiers is also 1.5 kΩ. With most 455 kHz ceramic filters and many crystal filters, no impedance matching network is necessary. The IF amplifier has 44 dB of gain and 5.5 MHz bandwidth. The IF limiter has 58 dB of gain and 4.5 MHz bandwidth.

To achieve optimum linearity of the log signal strength indicator, there must be a 12 dBV insertion loss between the first and second IF stages. If the IF filter or interstage network does not cause 12 dBV insertion loss, a fixed or variable resistor or an L pad for simultaneous loss and impedance matching can be added between the first IF output (IF_AMP_OUT) and the interstage network. The overall gain will then be 90 dB with 2 MHz bandwidth.

The signal from the second limiting amplifier goes to a Gilbert cell quadrature detector. One port of the Gilbert cell is internally driven by the IF. The other output of the IF is AC-coupled to a tuned quadrature network. This signal, which now has a 90° phase relationship to the internal signal, drives the other port of the multiplier cell.

The demodulated output of the quadrature drives an internal op amp. This op amp can be configured as a unity gain buffer, or for simultaneous gain, filtering, and second-order temperature compensation if needed. It can drive an AC load as low as 10 kΩ with a rail-to-rail output.

A log signal strength indicator completes the circuitry. The output range is greater than 70 dB and is temperature compensated. This signal drives an internal op amp. The op amp is capable of rail-to-rail output. It can be used for gain, filtering, or second-order temperature compensation of the RSSI, if needed.

Remark: dBV = 20log V_O/V_I.

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8. Limiting values

Table 3. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	supply voltage		-	7	V
T_{stg}	storage temperature		–65	+150	°C
T_{amb}	ambient temperature	operating	–40	+85	°C

9. Thermal characteristics

Table 4. Thermal characteristics

Symbol	Parameter	Conditions	Max	Unit
$Z_{th(j-a)}$	transient thermal impedance from junction to ambient	SA676DK/01 (SSOP20)	117	K/W

10. Static characteristics

Table 5. Static characteristics

V_{CC} = 3 V; T_{amb} = 25 °C; unless specified otherwise.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{CC}	supply voltage		2.7	-	7.0	V
I_{CC}	supply current		-	3.5	5.0	mA

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11. Dynamic characteristics

Table 6. Dynamic characteristics

T_{amb} = 25 °C; V_{CC} = 3 V; unless specified otherwise. RF frequency = 45 MHz + 14.5 dBV RF input step-up. IF frequency = 455 kHz; R17 = 2.4 kΩ and R18 = 3.3 kΩ. RF level = −45 dBm; FM modulation = 1 kHz with ±5 kHz peak deviation. Audio output with de-emphasis filter and C-message weighted filter. Test circuit Figure 9. The parameters listed below are tested using automatic test equipment to assure consistent electrical characteristics. The limits do not represent the ultimate performance limits of the device. Use of an optimized RF layout will improve many of the listed parameters.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Mixer/oscillator section (external LO = 220 mV RMS value)						
f_i	input frequency		-	100	-	MHz
f_{osc}	oscillator frequency		-	100	-	MHz
NF	noise figure	at 45 MHz	-	7.0	-	dB
$IP3_I$	input third-order intercept point	50 Ω source; f1 = 45.0 MHz; f2 = 45.06 MHz; input RF level = −52 dBm	-	−10	-	dBm
$G_{p(conv)}$	conversion power gain	matched 14.5 dBV step-up	10	17	-	dB
		50 Ω source	-	2.5	-	dB
$R_{i(RF)}$	RF input resistance	single-ended input	-	8	-	kΩ
$C_{i(RF)}$	RF input capacitance		-	3.0	4.0	pF
$R_{o(mix)}$	mixer output resistance	MIXER_OUT pin	1.25	1.5	-	kΩ
IF section						
$G_{amp(IF)}$	IF amplifier gain	50 Ω source	-	44	-	dB
G_{lim}	limiter gain	50 Ω source	-	58	-	dB
α_{AM}	AM rejection	30 % AM 1 kHz	-	50	-	dB
$V_{o(aud)}$	audio output voltage	gain of two	60	120	-	mV
SINAD	signal-to-noise-and-distortion ratio	IF level −110 dBm	-	17	-	dB
THD	total harmonic distortion		-	−55	-	dB
S/N	signal-to-noise ratio	no modulation for noise	-	60	-	dB
$V_{o(RSSI)}$	RSSI output voltage	IF; R9 = 2 kΩ	[1]			
		IF level = −110 dBm	-	0.5	0.9	V
		IF level = −50 dBm	-	1.7	2.2	V
$\alpha_{RSSI(range)}$	RSSI range		-	70	-	dB
$Z_{i(IF)}$	IF input impedance	IF_AMP_IN pin	1.3	1.5	-	kΩ
$Z_{o(IF)}$	IF output impedance	IF_AMP_OUT pin	-	0.3	-	kΩ
$Z_{i(lim)}$	limiter input impedance	LIMITER_IN pin	1.3	1.5	-	kΩ
$Z_{o(lim)}$	limiter output impedance	LIMITER_OUT pin	-	0.3	-	kΩ
$V_{o(RMS)}$	RMS output voltage	LIMITER_OUT pin	-	130	-	mV
RF/IF section (internal LO)						
SINAD	signal-to-noise-and-distortion ratio	system; RF level = −114 dBm	-	12	-	dB

[1] The generator source impedance is 50 Ω, but the SA676 input impedance at IF_AMP_IN (pin 18) is 1500 Ω. As a result, IF level refers to the actual signal that enters the SA676 input (IF_AMP_IN, pin 18), which is about 21 dB less than the 'available power' at the generator.

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12. Performance curves

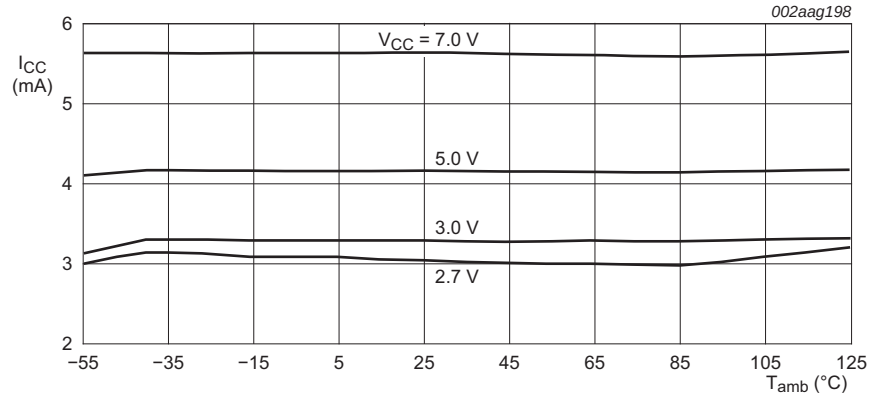


Fig 3. Supply current versus ambient temperature

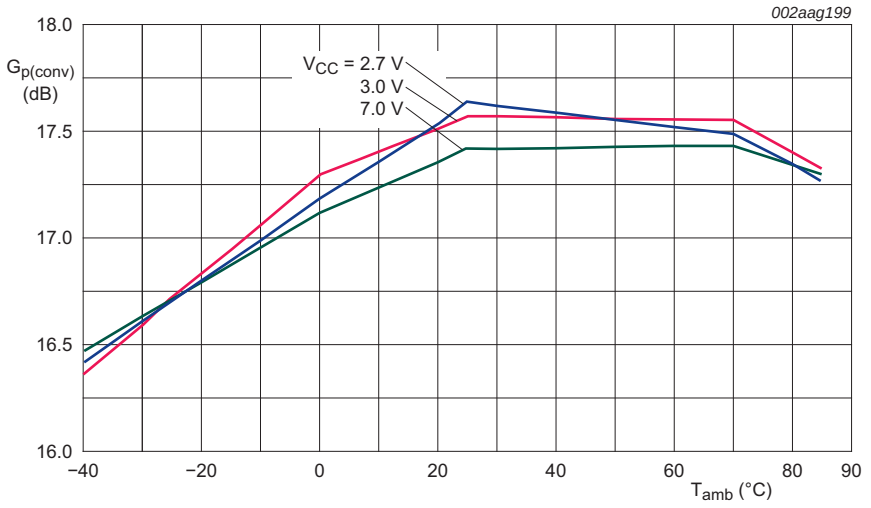
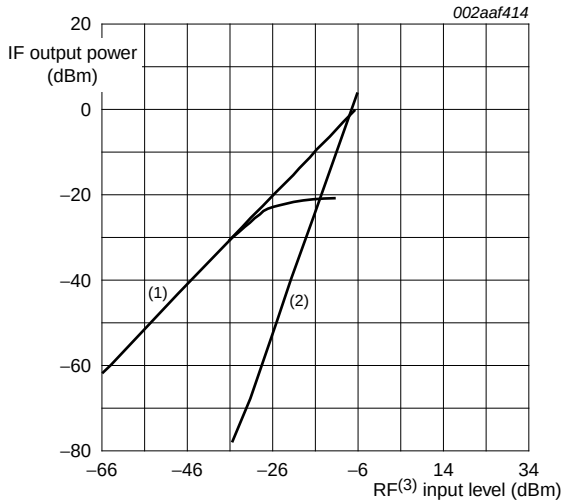


Fig 4. Conversion power gain versus ambient temperature

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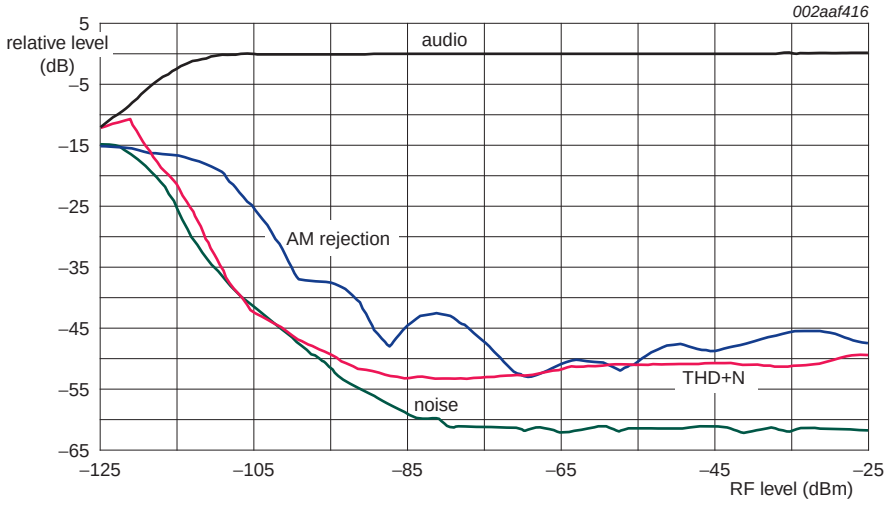
RF = 45 MHz; IF = 455 kHz.

(1) Fund product.

(2) Third order product.

(3) 50 Ω input.

Fig 5. Mixer third order intercept and compression

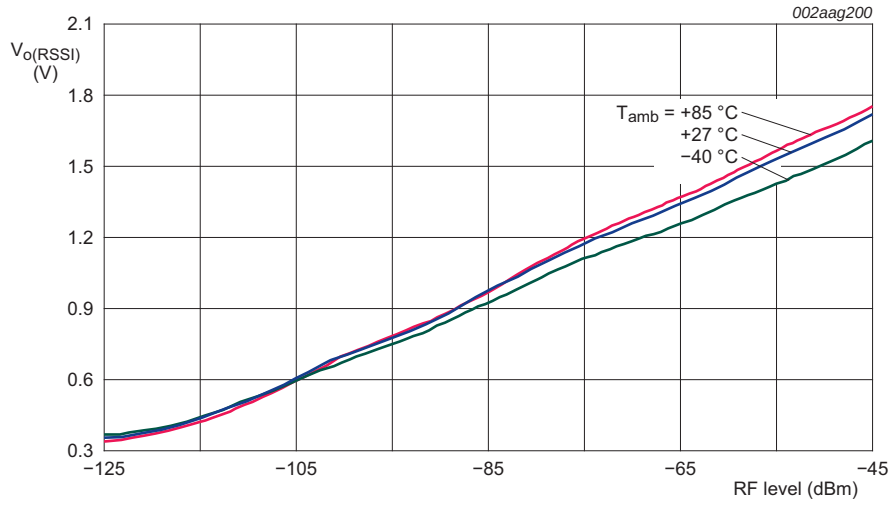


V_{CC} = 3 V; RF = 45 MHz; deviation = ±5 kHz; $V_{o(aud)RMS}$ = 117.6 mV.

Fig 6. Relative level of audio, AM rejection, THD+N and noise (T_{amb} = +25 °C)

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V_{CC} = 3 V

Fig 7. RSSI output voltage versus RF level

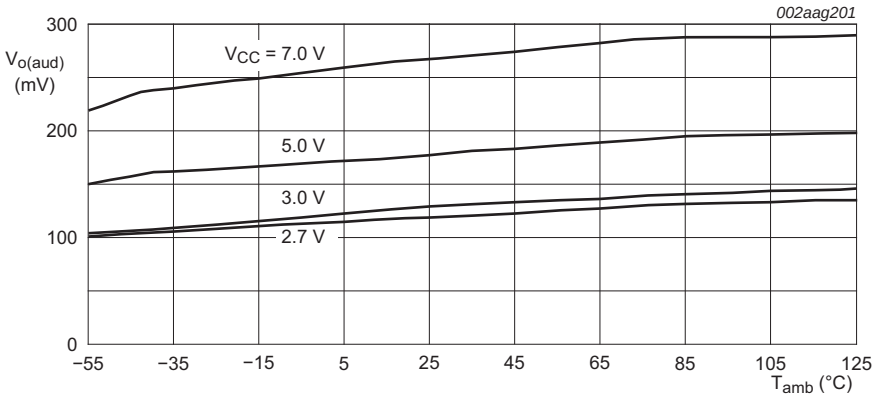
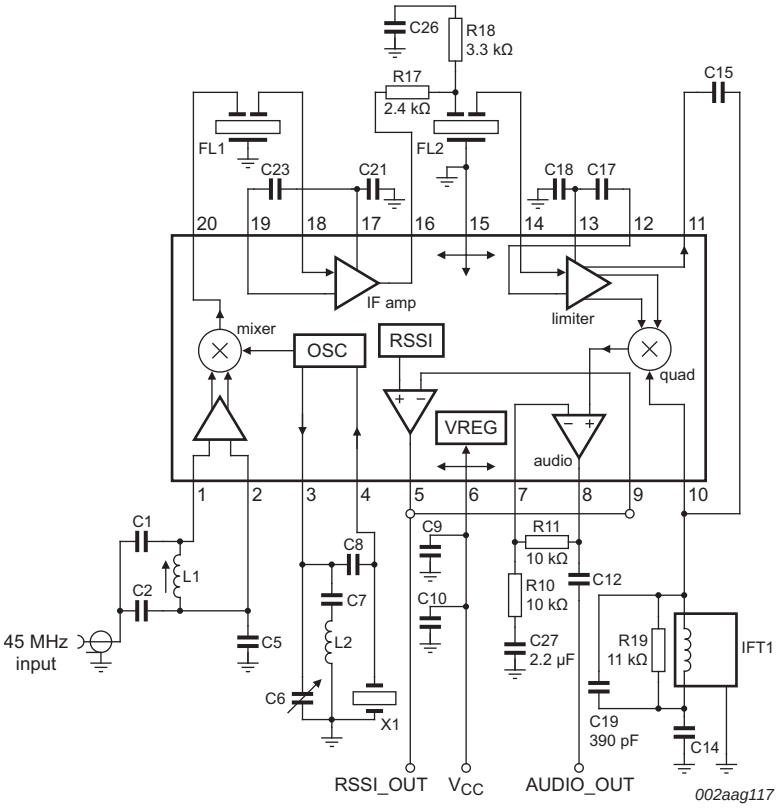


Fig 8. Audio output voltage versus ambient temperature

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13. Application information



The layout is very critical in the performance of the receiver. We highly recommend our demo board layout.

All of the inductors, the quad tank, and their shield must be grounded. A 10 µF to 15 µF or higher value tantalum capacitor on the supply line is essential. A low frequency ESR screening test on this capacitor will ensure consistent good sensitivity in production. A 0.1 µF bypass capacitor on the supply pin, and grounded near the 44.545 MHz oscillator improves sensitivity by 2 dB to 3 dB.

Fig 9. SA676 45 MHz application circuit (SA676DK demo board)

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Table 7. SA676DK demo board component list

Component	Description
C1	51 pF NPO ceramic
C2	220 pF NPO ceramic
C5, C9, C14, C17, C18, C21, C23, C26	100 nF ± 10 % monolithic ceramic
C6	5 pF to 30 pF trim cap
C7	1 nF ceramic
C8, C15	10.0 pF NPO ceramic
C10	10 μF tantalum (minimum)[1]
C12	2.2 μF ± 10 % tantalum
C19	390 pF ± 10 % monolithic ceramic
C27	2.2 μF tantalum
FL1, FL2[2]	ceramic filter Murata CFUKF455KB4X-R0
IFT1	330 μH Toko 303LN-1130
L1	330 nH Coilcraft UNI-10/142--04J08S
L2	0.8 μH nominal Toko 292CNS-T1038Z
X1	44.545 MHz crystal ICM4712701
R5[3]	not used in application board
R10	8.2 kΩ ± 5 % ¼ W carbon composition
R11	10 kΩ ± 5 % ¼ W carbon composition
R17	2.4 kΩ ± 5 % ¼ W carbon composition
R18	3.3 kΩ ± 5 % ¼ W carbon composition
R19	11 kΩ ± 5 % ¼ W carbon composition

[1] This value can be reduced when a battery is the power source.

[2] This is a 30 kHz bandwidth 455 kHz ceramic filter. All the characterization and testing are done with this wideband filter. A more narrowband 15 kHz bandwidth 455 kHz ceramic filter that may be used as an alternative selection is Murata CFUKG455KE4A-R0.

[3] R5 can be used to bias the oscillator transistor at a higher current for operation above 45 MHz. Recommended value is 22 kΩ, but should not be below 10 kΩ.

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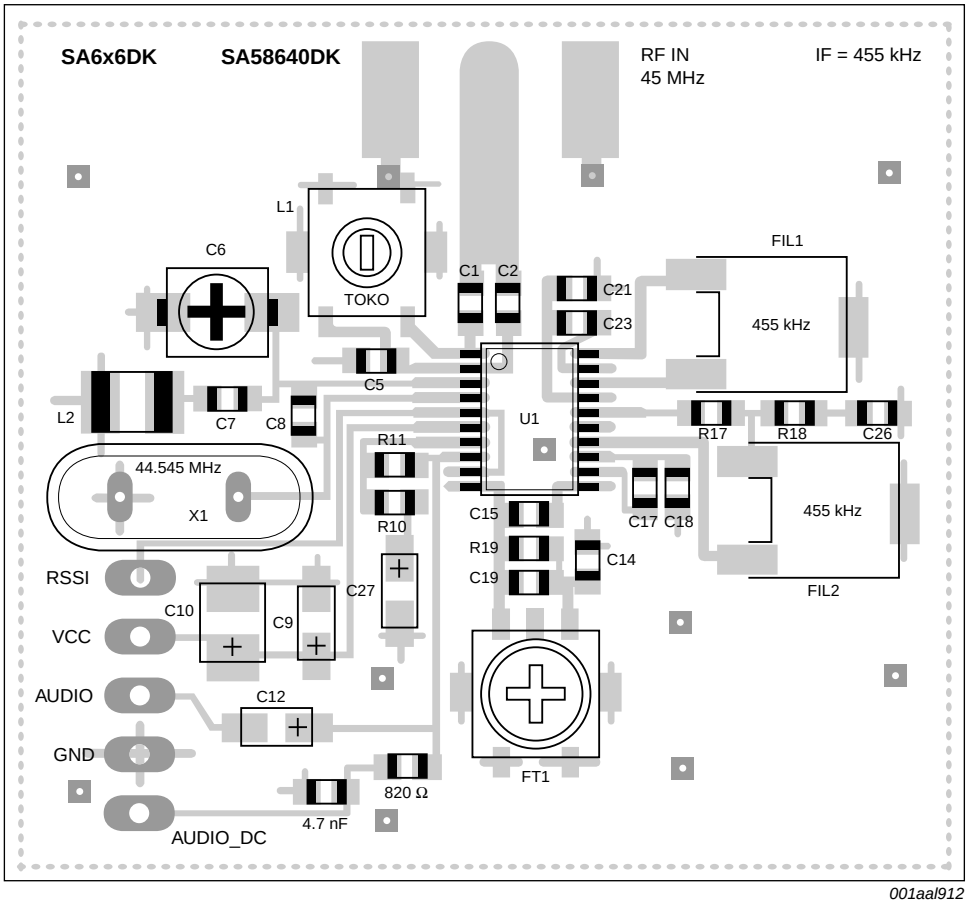


Fig 10. SA6x6DK/SA58640DK top view with components

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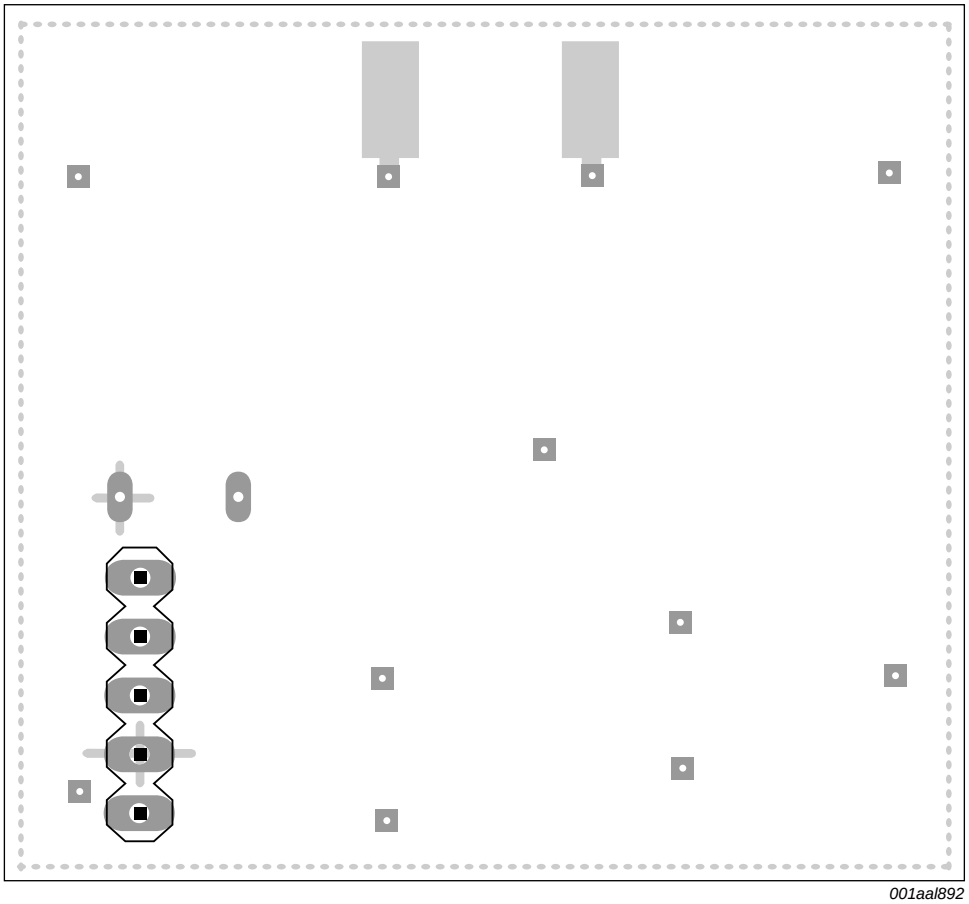
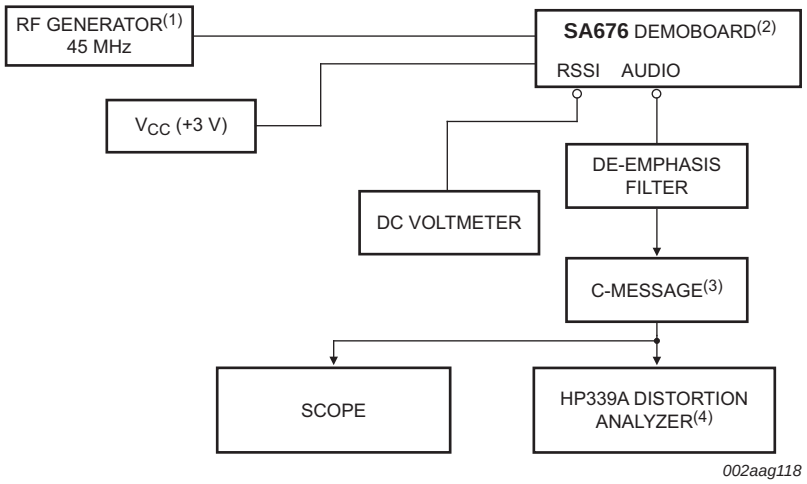


Fig 11. SA6x6DK/SA58640DK bottom view (viewed from top)

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14. Test information



(1) Set RF generator at 45.000 MHz; use a 1 kHz modulation frequency and a 6 kHz deviation if using 16 kHz filters, or 8 kHz if using 30 kHz filters.

(2) The smallest RSSI voltage (i.e., when no RF input is present and the input is terminated) is a measure of the quality of the layout and design. If the lowest RSSI voltage is 500 mV or higher, it means the receiver is in regenerative mode. In that case, the receiver sensitivity will be worse than expected.

(3) The C-message and de-emphasis filter combination has a peak gain of 10 dB for accurate measurements. Without the gain, the measurements may be affected by the noise of the scope and HP339A analyzer. The de-emphasis filter has a fixed −6 dB/octave slope between 300 Hz and 3 kHz.

(4) The measured typical sensitivity for 12 dB SINAD should be 0.45 μV or −114 dBm at the RF input.

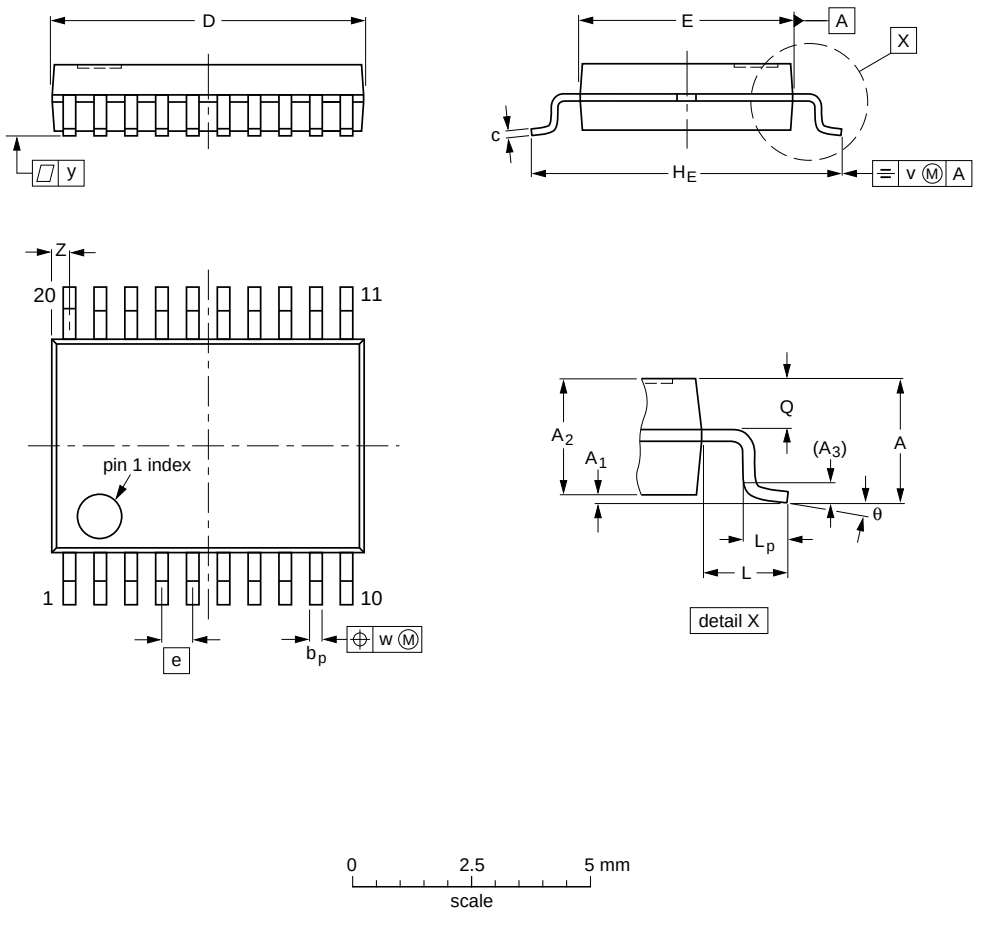
Fig 12. SA676 application circuit test setup

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15. Package outline

SSOP20: plastic shrink small outline package; 20 leads; body width 4.4 mm **SOT266-1**



DIMENSIONS (mm are the original dimensions)

UNIT	A max.	A_1	A_2	A_3	b_p	c	$D^{(1)}$	$E^{(1)}$	e	H_E	L	L_p	Q	v	w	y	$Z^{(1)}$	θ
mm	1.5	0.15 0	1.4 1.2	0.25	0.32 0.20	0.20 0.13	6.6 6.4	4.5 4.3	0.65	6.6 6.2	1	0.75 0.45	0.65 0.45	0.2	0.13	0.1	0.48 0.18	10^o 0^o

Note

1. Plastic or metal protrusions of 0.20 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	JEITA			
SOT266-1		MO-152				~~99-12-27~~ 03-02-19

Fig 13. Package outline SOT266-1 (SSOP20)

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16. Soldering of SMD packages

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note *AN10365 "Surface mount reflow soldering description"*.

16.1 Introduction to soldering

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

16.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- Through-hole components
- Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- Board specifications, including the board finish, solder masks and vias
- Package footprints, including solder thieves and orientation
- The moisture sensitivity level of the packages
- Package placement
- Inspection and repair
- Lead-free soldering versus SnPb soldering

16.3 Wave soldering

Key characteristics in wave soldering are:

- Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- Solder bath specifications, including temperature and impurities

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16.4 Reflow soldering

Key characteristics in reflow soldering are:

- Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see Figure 14) than a SnPb process, thus reducing the process window
- Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with Table 8 and 9

Table 8. SnPb eutectic process (from J-STD-020C)

Package thickness (mm)	Package reflow temperature (°C)	
	Volume (mm^3)	
	< 350	≥ 350
< 2.5	235	220
≥ 2.5	220	220

Table 9. Lead-free process (from J-STD-020C)

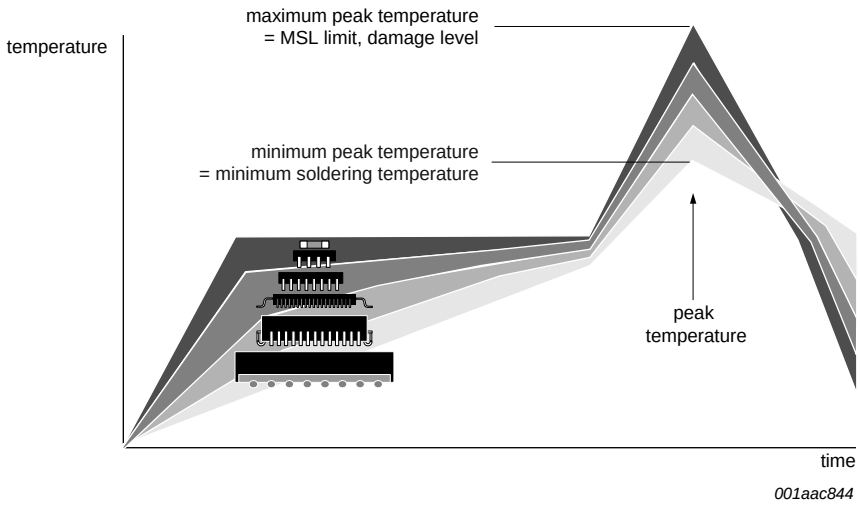
Package thickness (mm)	Package reflow temperature (°C)		
	Volume (mm^3)		
	< 350	350 to 2000	> 2000
< 1.6	260	260	260
1.6 to 2.5	260	250	245
> 2.5	250	245	245

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see Figure 14.

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MSL: Moisture Sensitivity Level

Fig 14. Temperature profiles for large and small components

For further information on temperature profiles, refer to Application Note *AN10365 "Surface mount reflow soldering description"*.

17. Abbreviations

Table 10. Abbreviations

Acronym	Description
AM	Amplitude Modulation
CDM	Charged-Device Model
ESD	ElectroStatic Discharge
ESR	Equivalent Series Resistance
FM	Frequency Modulation
HBM	Human Body Model
IF	Intermediate Frequency
LC	inductor-capacitor filter
LO	Local Oscillator
RF	Radio Frequency
RMS	Root Mean Squared
RSSI	Received Signal Strength Indicator

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18. Revision history

Table 11. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
SA676 v.3	20120719	Product data sheet	-	SA676 v.2
Modifications:	• Section 2 “Features and benefits”: – 13th bullet item re-written – added (new) 14th bullet item			
SA676 v.2	20110412	Product data sheet	-	SA676 v.1
SA676 v.1	19931215	Product specification	ECN 853-1726 11659	-

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19. Legal information

19.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL http://www.nxp.com.

19.2 Definitions

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21. Contents

1	**General description**	**1**
2	**Features and benefits**	**1**
3	**Applications**	**1**
4	**Ordering information**	**2**
5	**Block diagram**	**2**
6	**Pinning information**	**3**
6.1	Pinning	3
6.2	Pin description	3
7	**Functional description**	**4**
8	**Limiting values**	**5**
9	**Thermal characteristics**	**5**
10	**Static characteristics**	**5**
11	**Dynamic characteristics**	**6**
12	**Performance curves**	**7**
13	**Application information**	**10**
14	**Test information**	**14**
15	**Package outline**	**15**
16	**Soldering of SMD packages**	**16**
16.1	Introduction to soldering	16
16.2	Wave and reflow soldering	16
16.3	Wave soldering	16
16.4	Reflow soldering	17
17	**Abbreviations**	**18**
18	**Revision history**	**19**
19	**Legal information**	**20**
19.1	Data sheet status	20
19.2	Definitions	20
19.3	Disclaimers	20
19.4	Trademarks	21
20	**Contact information**	**21**
21	**Contents**	**22**

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